

CMX90G701 6 - 18 GHz Positive Gain-Slope Amplifier +1dB

Description

The CMX90G701 is a low-power 50 Ω gain block suitable for a wide variety of wireless applications operating in the 6 – 18 GHz frequency range.

The gain block has a positive gain-slope of +1dB across the 6 – 16 GHz band, eliminating the need for equalisation and compensates for increasing system losses with frequency.

CMX90G701 is highly integrated for ease of use, minimising external component count and board area. RF ports are matched on-chip to 50 Ω with an output DC-blocking capacitor. An active bias circuit with enable function, allows the device to operate over a wide supply voltage of 2 V to 5 V with typical current of 22 mA.

The device is fabricated using a GaAs pHEMT process to provide a combination of high linearity, low noise and low DC power consumption.

An alternative part, CMX90G702, is available for applications that require more gain-slope compensation.

Applications

- Satcom Ku-band
- Fixed wireless access (FWA)
- 5G infrastructure & backhaul
- X-band (8 – 12GHz)
- PA Modules
- Eliminate passive equaliser

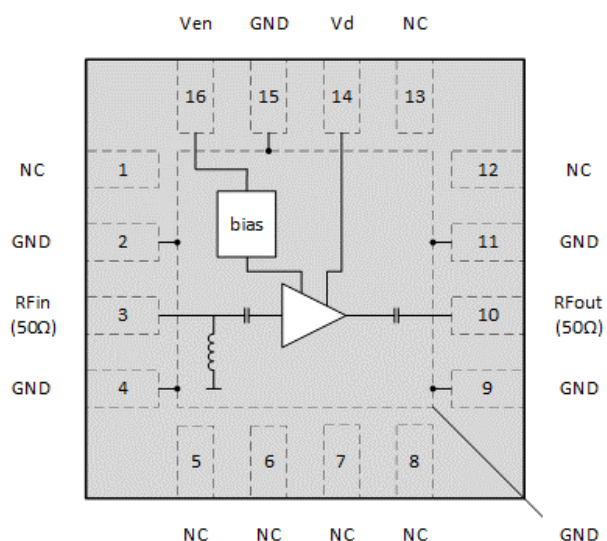


3x3mm VQFN-16 Package

Product Features

- Wide frequency range 6 – 18 GHz
- Positive gain-slope +1 dB (6 – 16 GHz)
- Small signal gain 9.4 – 10.4 dB
- Single positive DC supply 2 – 5 V
- Low current consumption (typ 22 mA)
- Low noise figure 3 dB
- Output P1dB +10 dBm @ 12 GHz
- 1.8 V logic compatible enable
- 105°C operating temperature ($V_d \leq 3.3$ V)

Block Diagram



Ordering Information

Part Number	Description
CMX90G701QF-R705	7" Reel with 500 pieces
CMX90G701QF-R710	7" Reel with 1,000 pieces
EV90G701	Evaluation board

Absolute Maximum Ratings

Parameter	Rating
RF Input Power	+12dBm
Device Voltage (Vd, Ven)	+5.5V
Case Temperature (Tc)	-40 to +85 °C (Vd > 3.3V), -40 to +105 °C (Vd ≤ 3.3V)
Junction Temperature (Tjmax)	165 °C (Process MTTF = 10 ⁶ hours)
Storage Temperature	-40 to +125 °C
ESD Sensitivity	HBM 250V (Class 1A), CDM 750V (Class C2b)
MSL Level	Level 3

Exceeding the maximum ratings may result in damage or reduced device reliability.

Thermal Characteristics

Parameter	Rating
Thermal Resistance (Rjc)	540 °C/W (Tc = 85 °C, Vd = Ven = 5 V) 550 °C/W (Tc = 105 °C, Vd = Ven = 3.3V)

Thermal resistance is junction-to-case, where case refers to the exposed die pad on the backside which is in contact with the board.

Recommended Operating Conditions

Parameter	Min	Typ	Max	Units
Operating Frequency Range	6		18	GHz
Case Temperature (Tc) Vd > 3.3V	-40		+85	°C
Case Temperature (Tc) Vd ≤ 3.3V	-40		+105	°C
Device Voltage (Vd)	2		5	V
Enable Voltage (Ven)	0		5	V

The device will be tested under certain conditions, but performance is not guaranteed over the full range of recommended operating conditions.

ESD Caution



CMX90G701 incorporates ESD protection circuitry however ESD precautions are strongly recommended for handling and assembly. Ensure that devices are protected from ESD in antistatic bags or carriers when being transported. Personal grounding is to be worn at all times when handling these devices.

RoHS Compliance



All devices supplied by CML Microcircuits are compliant with RoHS directive (2011/65/EU), containing less than the permitted levels of hazardous substances

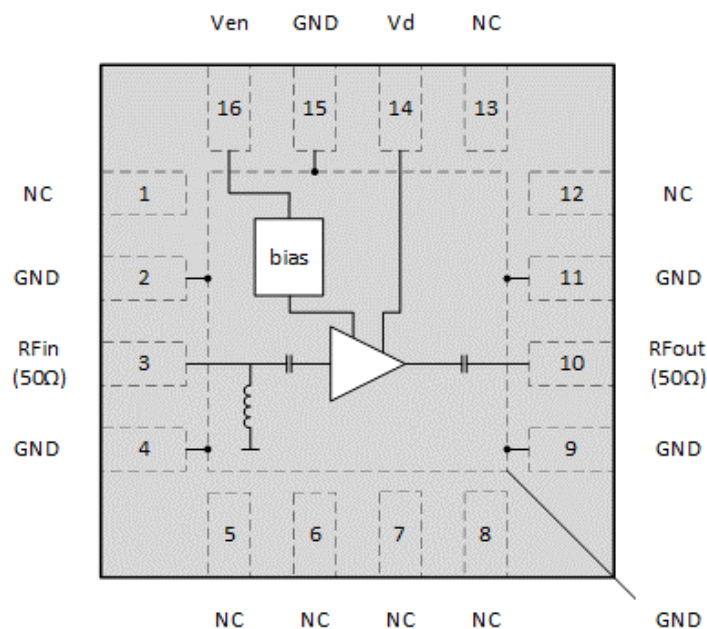
Electrical Specification

Results taken on EV90G701 EVB, where track losses have been de-embedded using the calibration line on the EV90G701 evaluation board.

$Z_0 = 50 \Omega$, $V_d = +5 \text{ V}$, $V_{en} = +5 \text{ V}$, $T_a = +25^\circ \text{C}$ (unless otherwise noted)

Parameter	Conditions	Min	Typ	Max	Units
Frequency		6		18	GHz
Small Signal Gain	6 GHz		9.4		dB
Small Signal Gain	16 GHz		10.4		dB
Small Signal Gain	18 GHz		10.2		dB
Gain Slope	6 GHz to 16 GHz		+1		dB
Reverse Isolation	6 GHz to 18 GHz		>16		dB
P1dB	At 12 GHz		10		dBm
OIP3	Two-tone test $\Delta f = 10 \text{ MHz}$, at 12 GHz, Pout/Tone = 0dBm	-	19	-	dBm
Noise Figure	12GHz		2.6		dB
Input Return Loss	6 GHz to 18 GHz		>10		dB
Output Return Loss	6 GHz to 18 GHz		>10		dB
Device Current (Id)			22		mA
Ven (Logic 1 = Enabled)	Amplifier normal operation	1.8		5	V
Ven (Logic 0 = Standby)	Amplifier in standby mode	0		0.2	V
Id	Ven = 0 V		1.5		uA
Id	Ven = 0.2 V		15		uA
S21	Ven = 0 to 0.2 V		-11.8		dB
Ven Current (Ien)	Ven = 5 V		0.8		mA
Turn-On Time RFout: 10 % to 90 %	RFin = -10 dBm, 10 GHz		0.6		μs
Turn-Off Time RFout: 90 % to 10 %	RFin = -10 dBm, 10 GHz		0.9		μs

Pin Assignments



Top View

Pin	Name	Description
1	NC	Connect to GND
2	GND	Connect to GND
3	RFin	RF input. Internally matched to 50 Ω with DC path to ground to provide enhanced ESD robustness.
4	GND	Connect to GND
5	NC	Connect to GND
6	NC	Connect to GND
7	NC	Connect to GND
8	NC	Connect to GND
9	GND	Connect to GND
10	RFout	RF output. Internally matched to 50 Ω with integrated DC-blocking capacitor.
11	GND	Connect to GND
12	NC	Connect to GND
13	NC	Connect to GND
14	Vd	Voltage supply to amplifier
15	GND	Connect to GND
16	Ven	Amplifier enable input
Die pad	GND	DC and RF ground. Exposed die pad must be connected to GND.

Notes

CML recommends that all no connect (NC) pins are connected to ground.

The bottom exposed die pad must be connected to the ground plane on the board.

Typical Performance

The following plots show typical performance characteristics of CMX90G701 measured on the evaluation board (Part Number EV90G701). Board losses have been de-embedded from the measurement results using the through line that is included on the EV90G701.

RF typical performance: $V_d = V_{en} = 5.0V$, $T_a = 25^\circ C$, $Z_0 = 50\ \Omega$

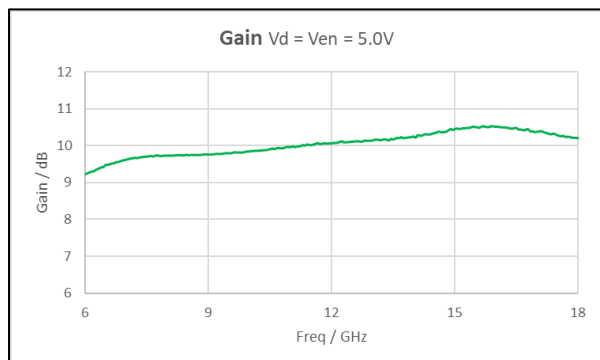


Figure 1: Gain $V_d = V_{en} = 5.0V$

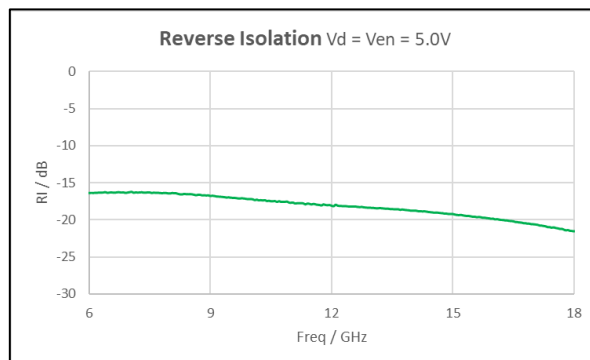


Figure 2: Reverse Isolation $V_d = V_{en} = 5.0V$

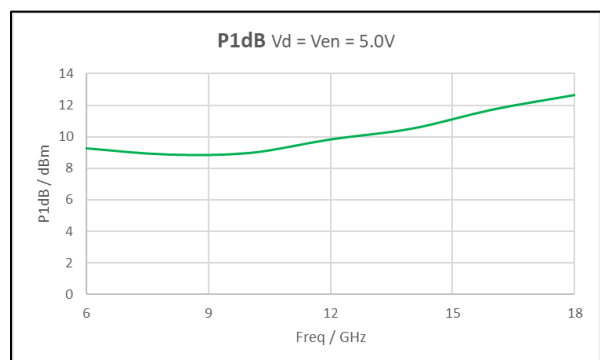


Figure 3: P1dB $V_d = V_{en} = 5.0V$

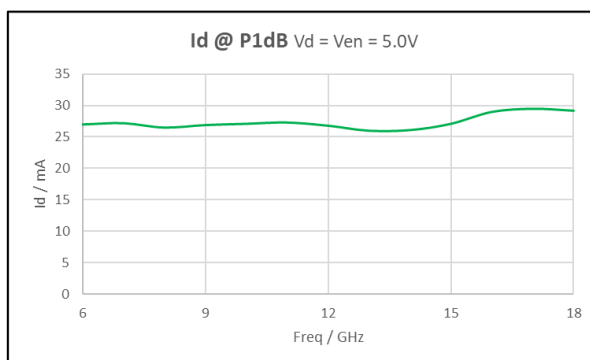


Figure 4: Id @ P1dB $V_d = V_{en} = 5.0V$

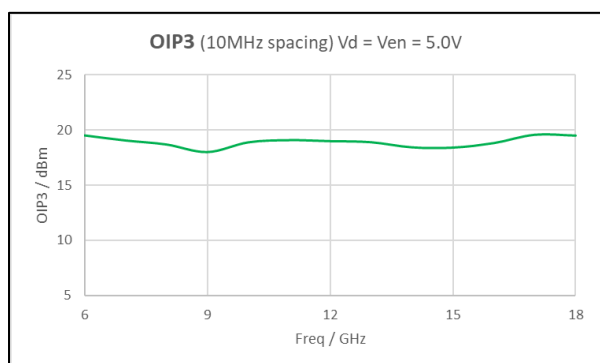


Figure 5: OIP3 (10MHz spacing) $V_d = V_{en} = 5.0V$

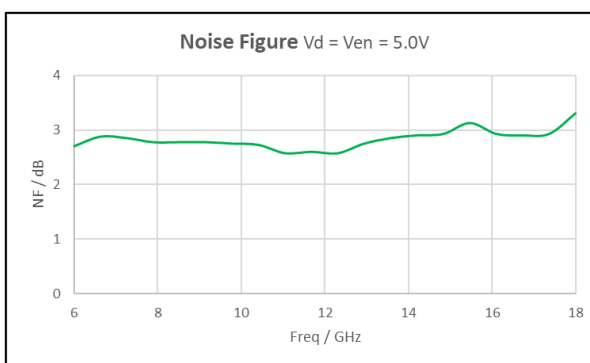


Figure 6: Noise Figure $V_d = V_{en} = 5.0V$

RF typical performance: $V_d = V_{en} = 5.0V$, $T_a = 25^\circ C$, $Z_0 = 50\ \Omega$ (continued)

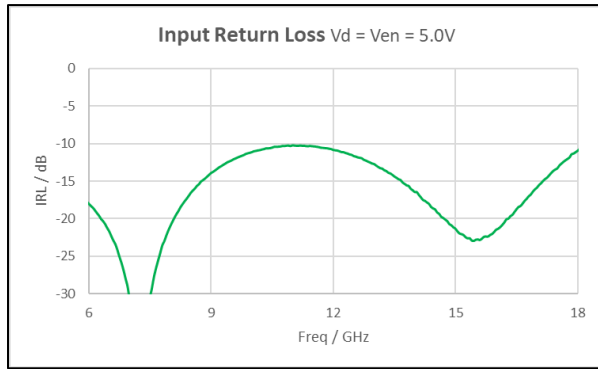


Figure 7: Input Return Loss $V_d = V_{en} = 5.0V$

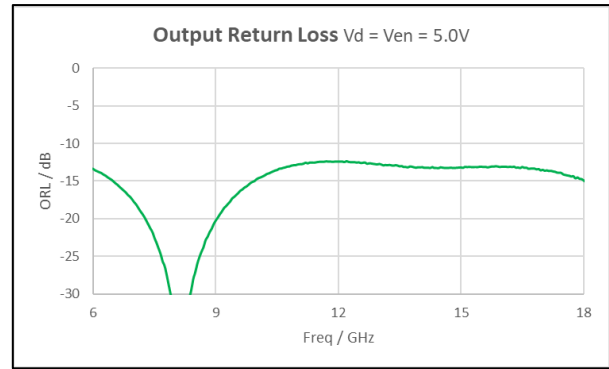


Figure 8: Output Return Loss $V_d = V_{en} = 5.0V$

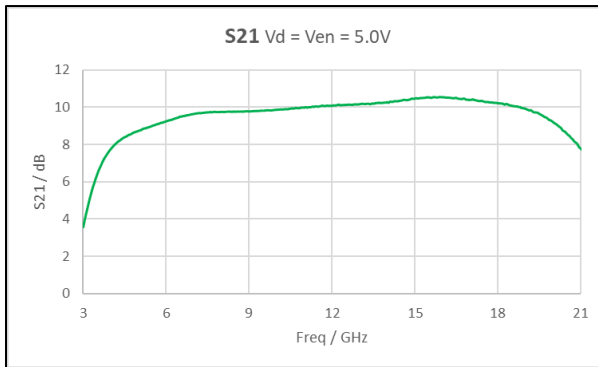


Figure 9: S_{21} $V_d = V_{en} = 5.0V$

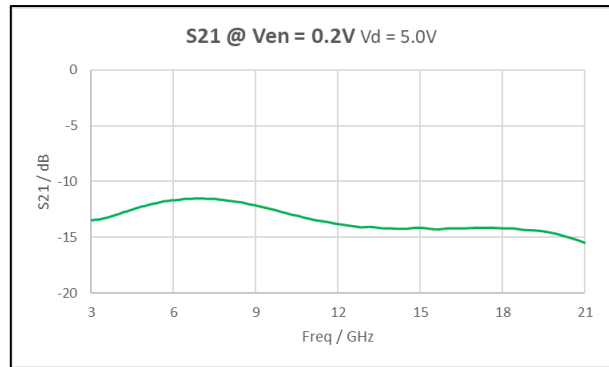


Figure 10: S_{21} @ $V_{en} = 0.2V$ $V_d = 5.0V$

DC typical performance: $V_d = 5.0\ V$, $T_a = 25^\circ C$

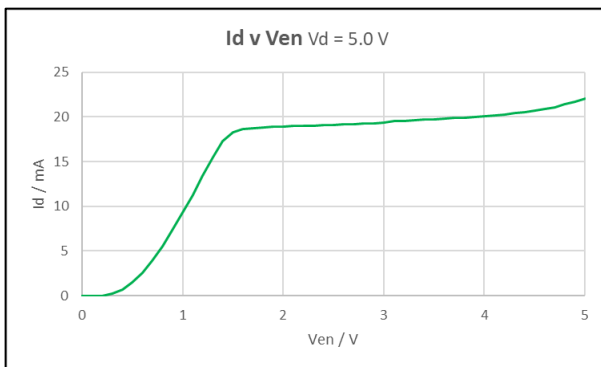


Figure 11: I_d v V_{en} $V_d = 5.0V$

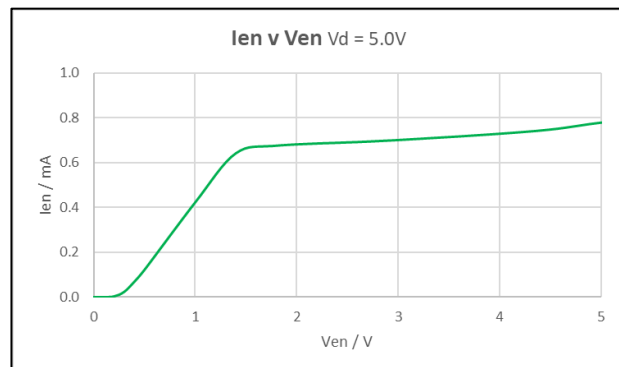
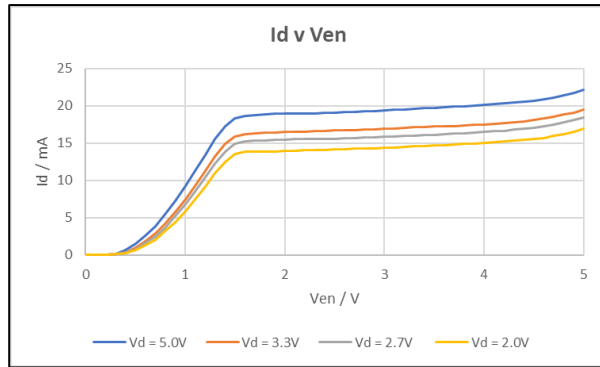
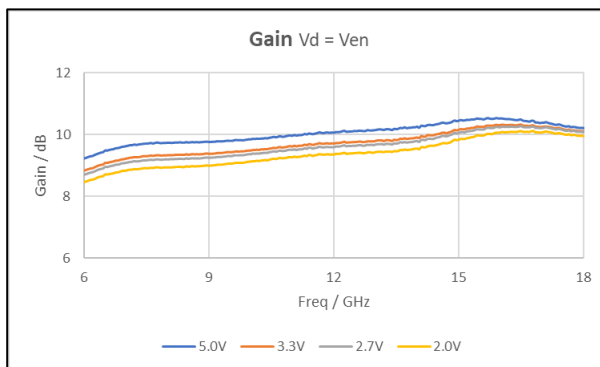
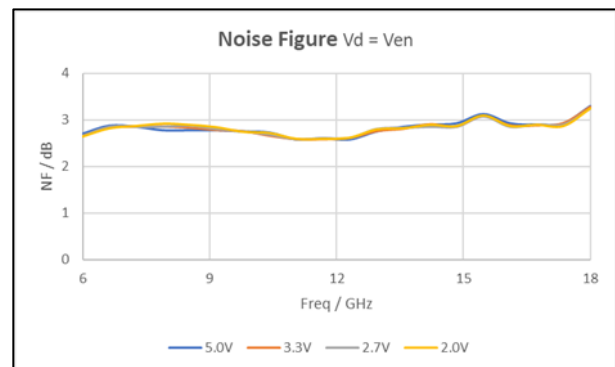
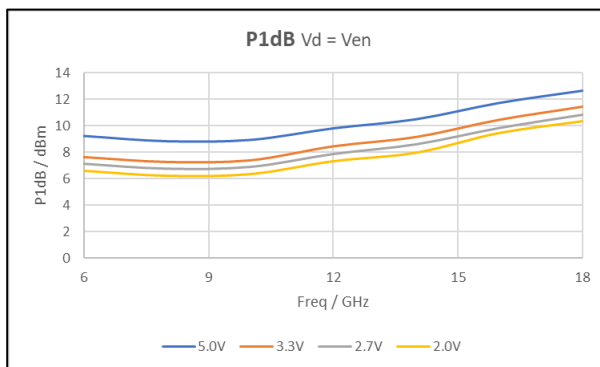
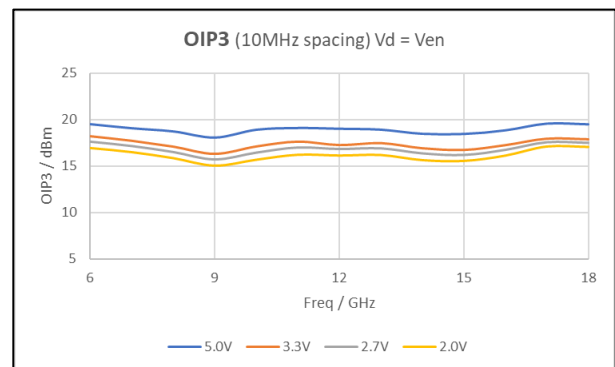
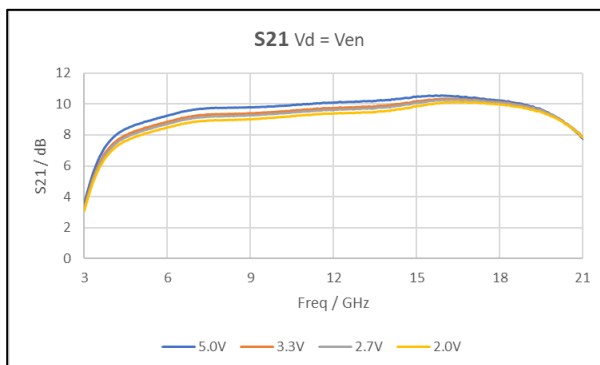
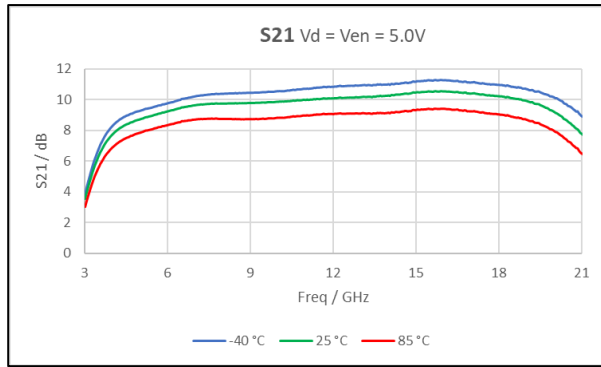
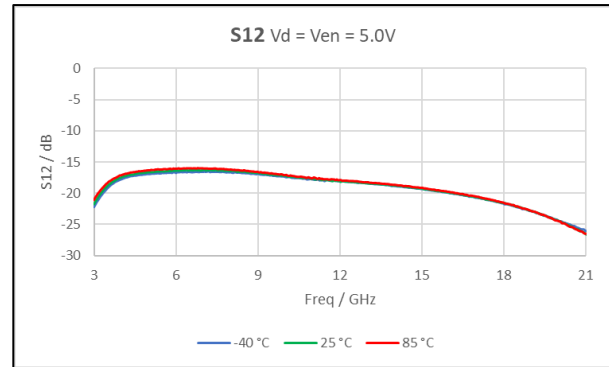
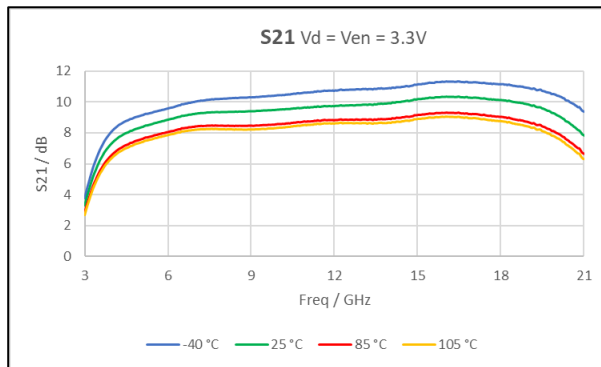
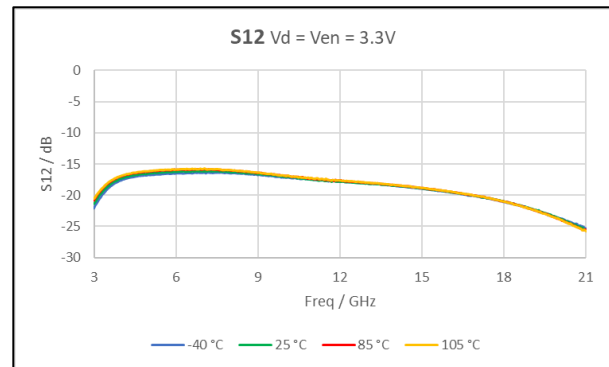
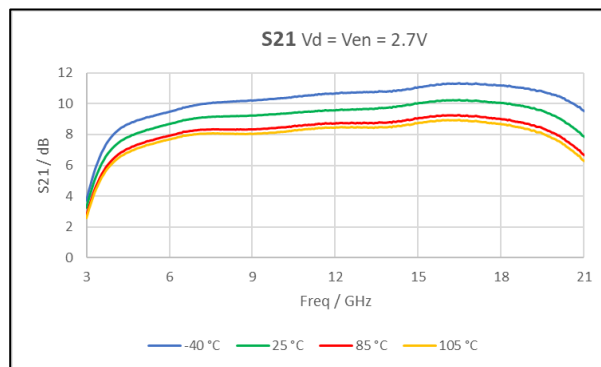
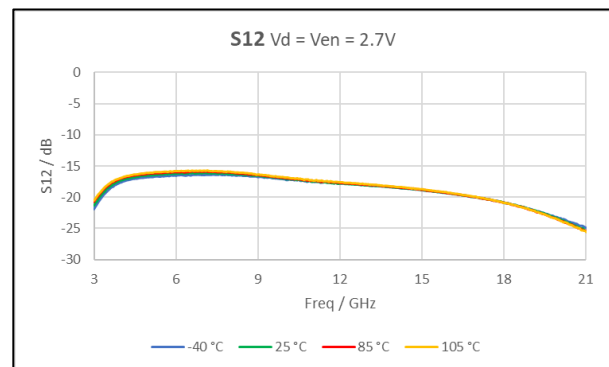
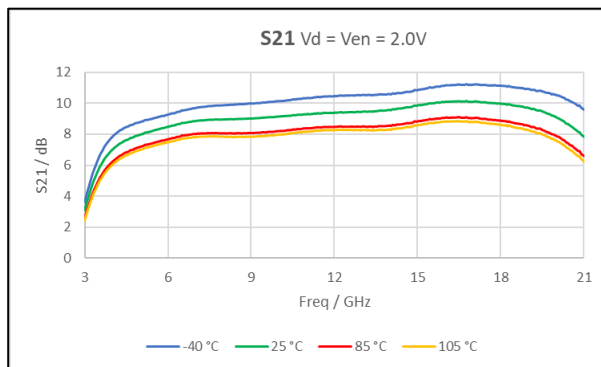
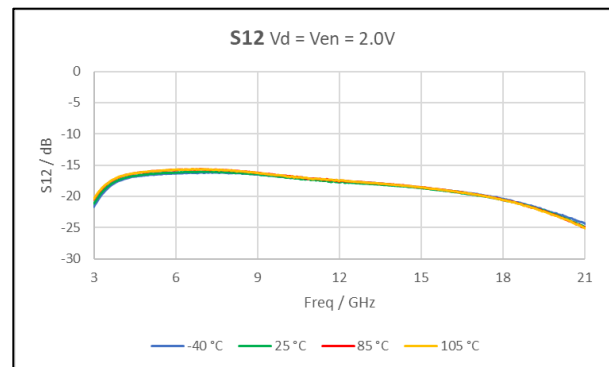


Figure 12: I_{en} v V_{en} $V_d = 5.0V$

DC Performance over voltage: $T_a = 25^\circ\text{C}$ **Figure 13: Id v Ven****RF Performance over voltage: $V_d = V_{en}$, $T_a = 25^\circ\text{C}$, $Z_0 = 50\Omega$** **Figure 14: Gain Vd = Ven****Figure 15: Noise Figure Vd = Ven****Figure 16: P1dB Vd = Ven****Figure 17: OIP3 (10MHz spacing) Vd = Ven****Figure 18: S21 Vd = Ven**

RF Performance over temperature: $V_d = V_{en}$, $Z_0 = 50\Omega$ Figure 19: S21 $V_d = V_{en} = 5.0V$ Figure 20: S12 $V_d = V_{en} = 5.0V$ Figure 21: S21 $V_d = V_{en} = 3.3V$ Figure 22: S12 $V_d = V_{en} = 3.3V$ Figure 23: S21 $V_d = V_{en} = 2.7V$ Figure 24: S12 $V_d = V_{en} = 2.7V$ Figure 25: S21 $V_d = V_{en} = 2.0V$ Figure 26: S12 $V_d = V_{en} = 2.0V$

RF performance over temperature: $V_d = V_{en}$, $Z_0 = 50\ \Omega$

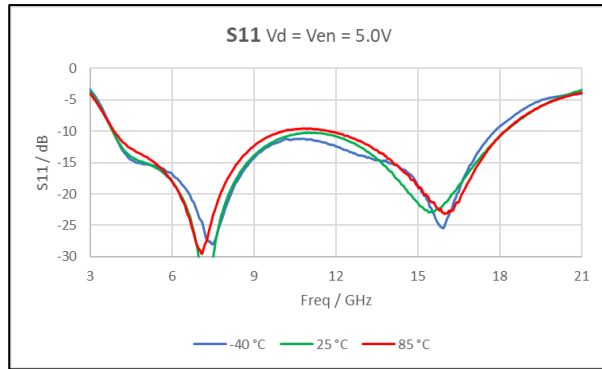


Figure 27: S11 $V_d = V_{en} = 5.0V$

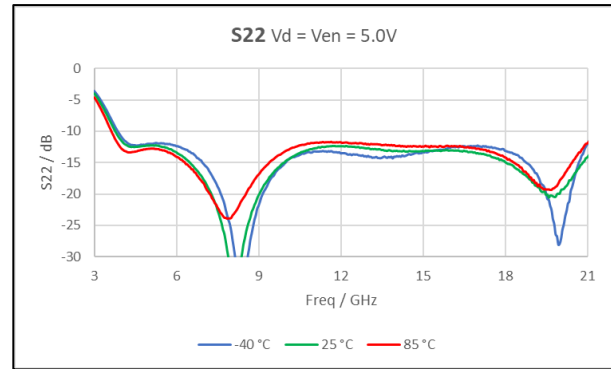


Figure 28: S22 $V_d = V_{en} = 5.0V$

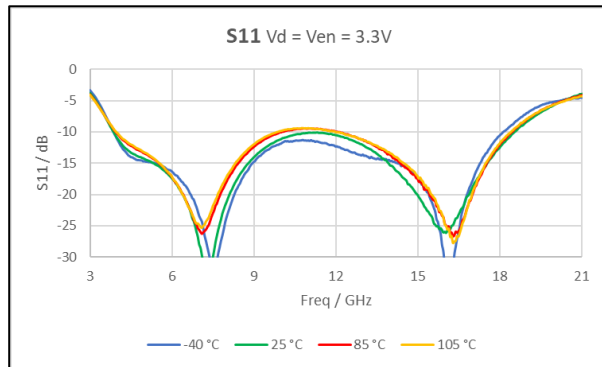


Figure 29: S11 $V_d = V_{en} = 3.3V$

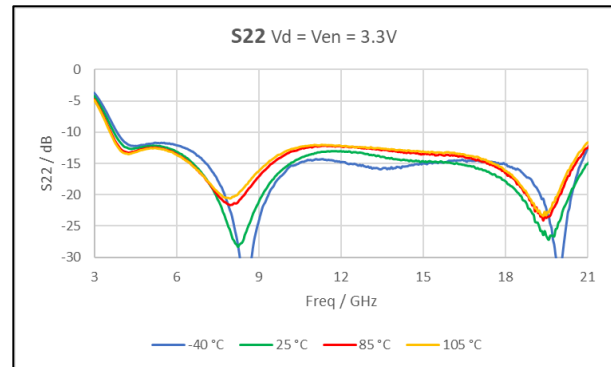


Figure 30: S22 $V_d = V_{en} = 3.3V$

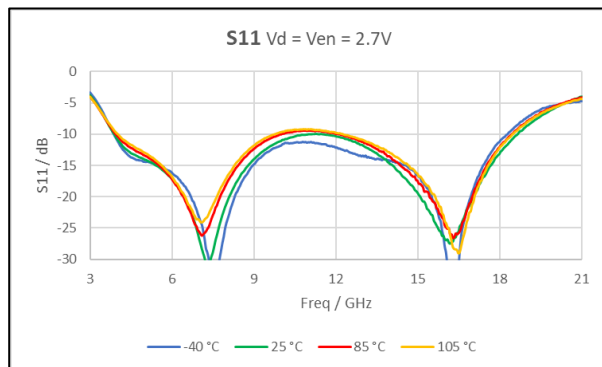


Figure 31: S11 $V_d = V_{en} = 2.7V$

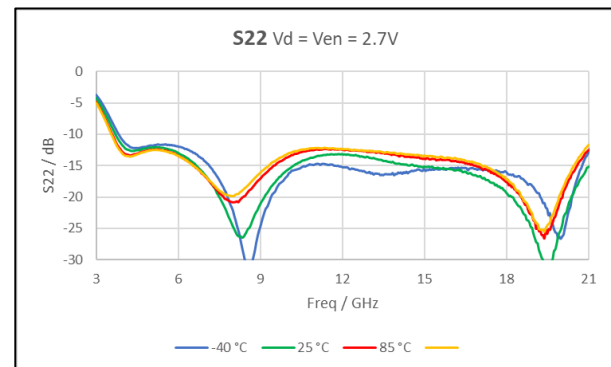


Figure 32: S22 $V_d = V_{en} = 2.7V$

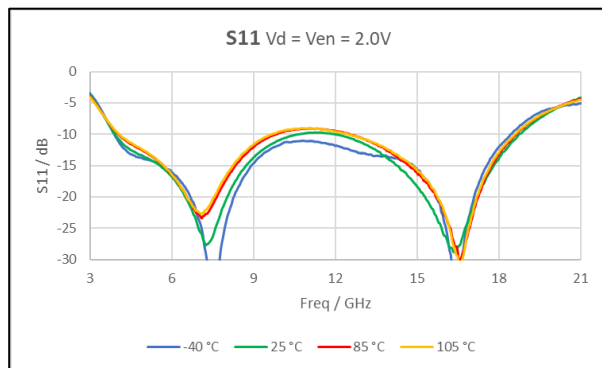


Figure 33: S11 $V_d = V_{en} = 2.0V$

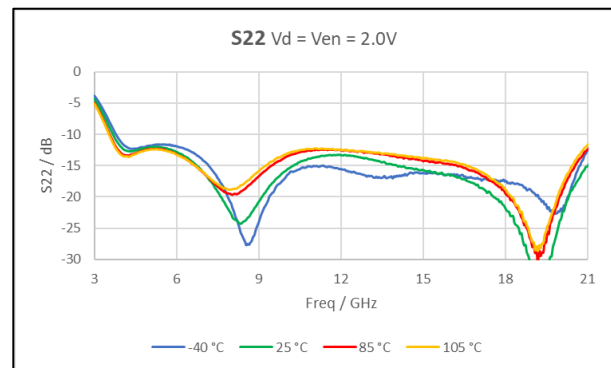


Figure 34: S22 $V_d = V_{en} = 2.0V$

RF performance over temperature: $V_d = V_{en}$, $Z_0 = 50\ \Omega$

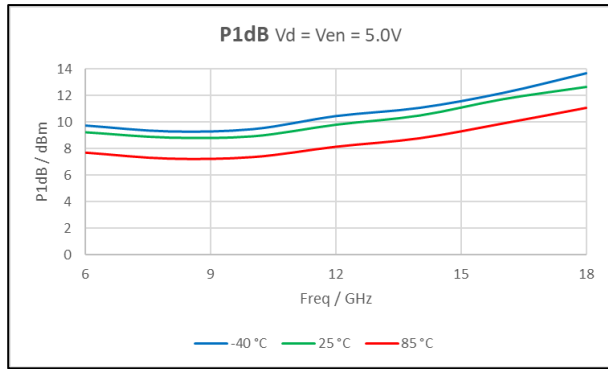


Figure 35: P1dB $V_d = V_{en} = 5.0V$

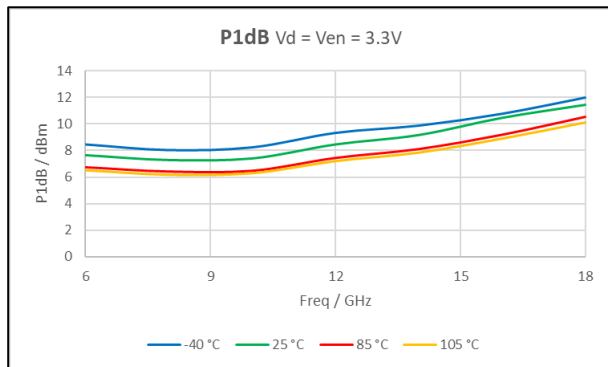


Figure 36: P1dB $V_d = V_{en} = 3.3V$

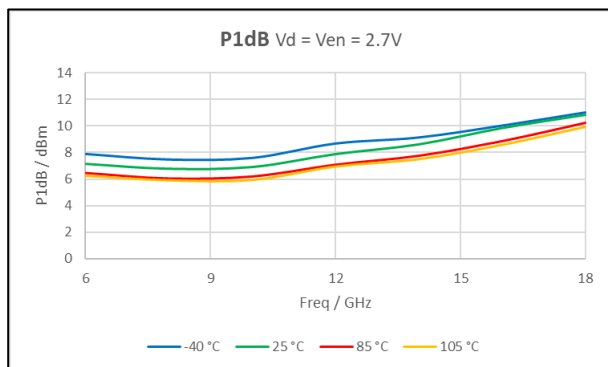


Figure 37: P1dB $V_d = V_{en} = 2.7V$

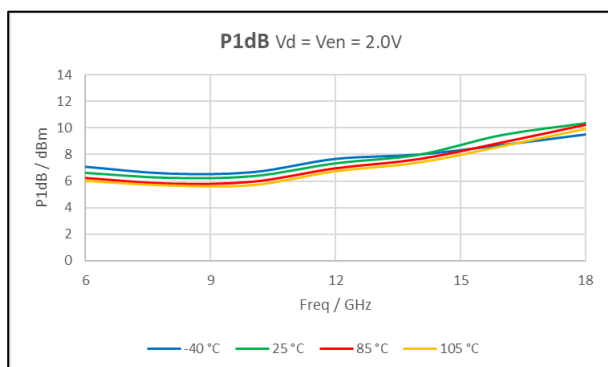
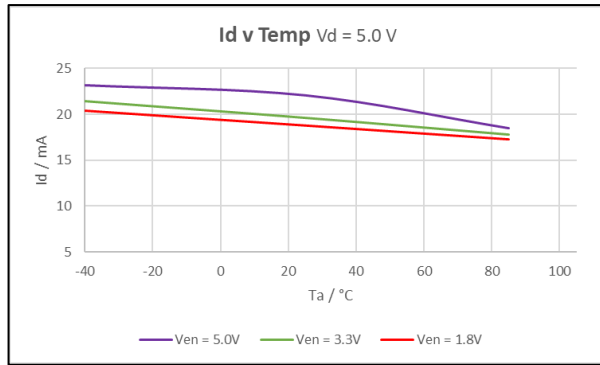
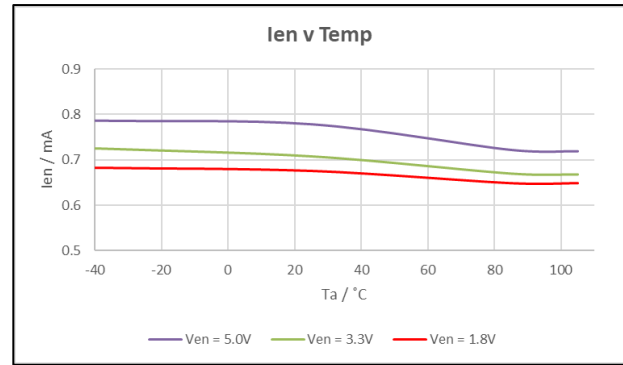
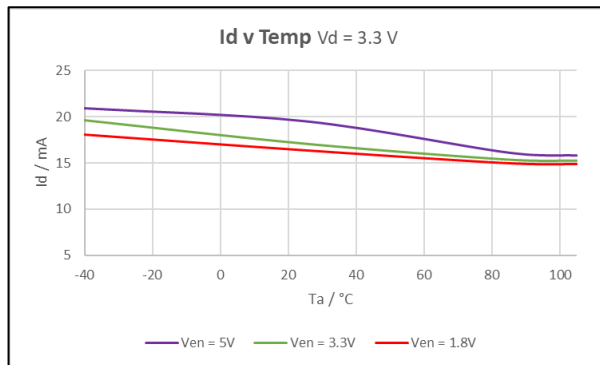
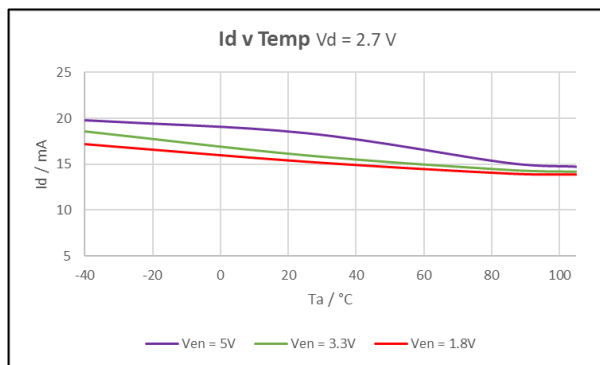
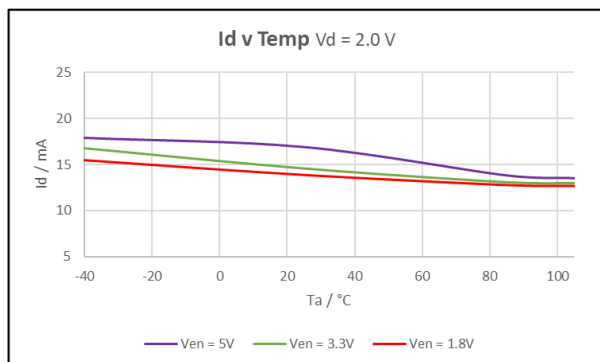


Figure 38: P1dB $V_d = V_{en} = 2.0V$

DC performance over temperature

Figure 39: I_d v Temp $V_d = 5.0$ VFigure 40: I_{en} v TempFigure 41: I_d v Temp $V_d = 3.3$ VFigure 42: I_d v Temp $V_d = 2.7$ VFigure 43: I_d v Temp $V_d = 2.0$ V

Application Information

Schematic Diagram

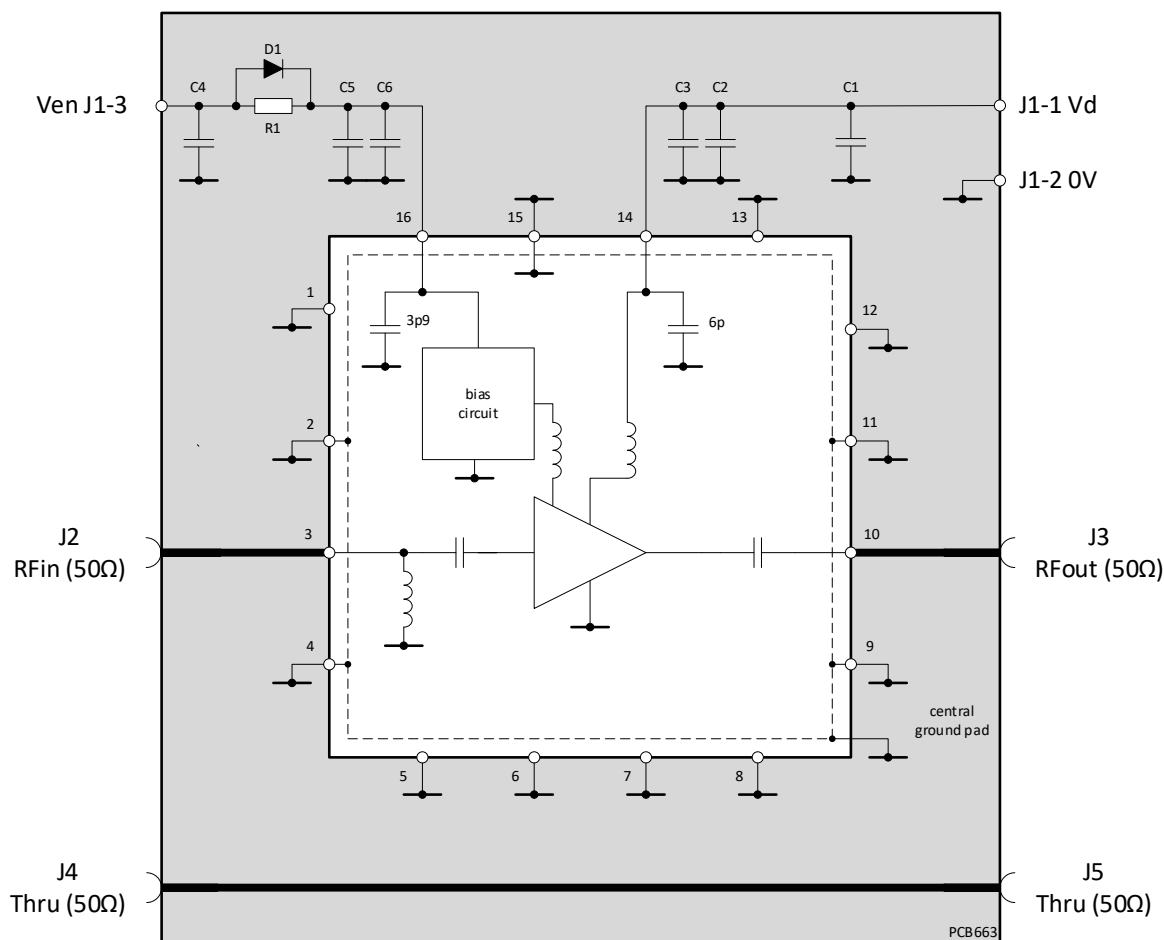


Figure 44: EV90G701 Schematic

Bill Of Materials (BOM)

Reference Designator	Value	Size	Description
C1	1 uF	0603	16 V, +/- 10 %
C2	10 nF	0402	16 V, +/- 10%
C3	DNF	0402	
C4	10 nF	0402	16 V, +/- 10%
C5	1 nF	0402	25 V, +/- 5%
C6	DNF	0402	
R1	0 R	0402	0.063 W
D1	DNF	SOD-523F	

Notes

- DNF = Do not fit component

PCB Layout

Careful layout of the printed circuit board (PCB) is essential for optimum RF and thermal performance. The recommended layout, including ground via pattern underneath the device, may be taken from the evaluation board (Part Number EV90G701).

The PCB consists of a top layer of MT40 backed by 2 layers of FR-4 with a total thickness of 1.632 mm (Figure 45) and the EV90G701 PCB (Figure 46) is 20 mm x 45 mm. The coplanar RF transmission lines have a width of 0.33 mm with a gap of 0.14 mm to ground either side. The through line length has been reduced by 3mm to account for the length of the device.

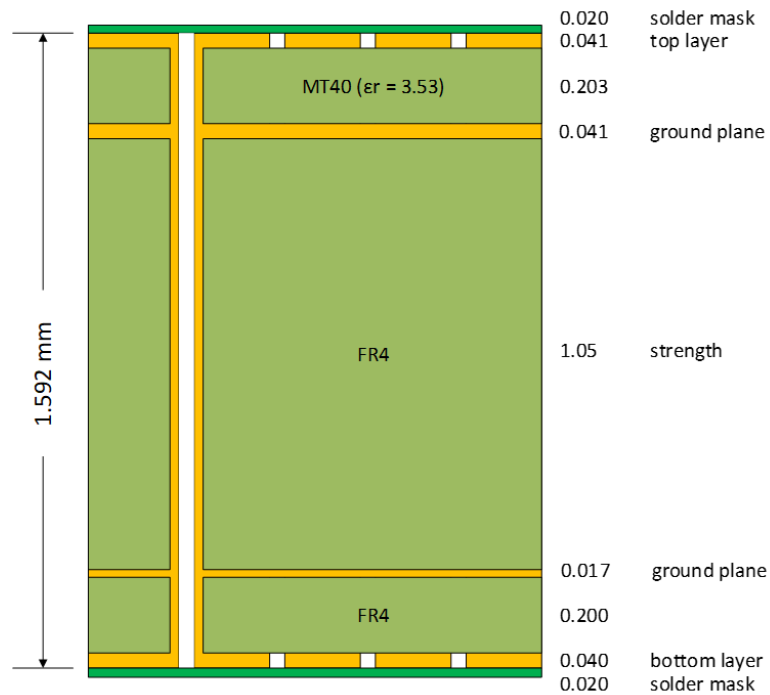


Figure 45: EV90G701 Layer Stack

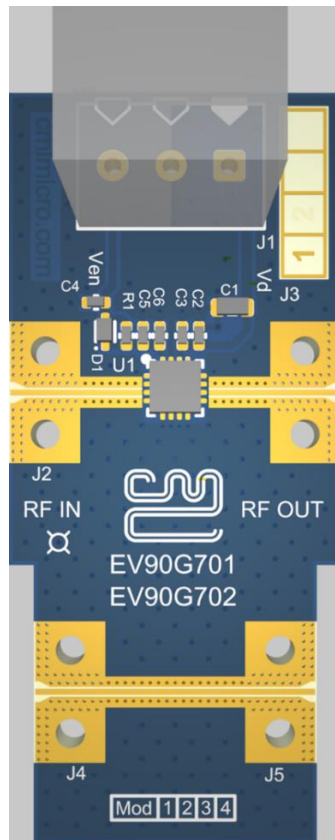


Figure 46: EV90G701 PCB Top Layer View

Thermal Design

The primary RF/DC ground and thermal path is via the exposed die pad on the backside of the package, which must be connected to the PCB ground plane. An array of plated through-hole vias directly underneath the die pad area is essential to conduct heat away and minimise ground inductance. A typical solution will have 9 grounding vias connecting the top layer to the bottom layer, with inner diameter of 0.2 mm (and 0.025 mm plating) on a 0.55 mm grid pattern. The vias do not need to be filled. The PCB layout should provide a thermal radiator appropriate for the intended operation, adding as much copper to inner and outer layers as possible to avoid excessive junction temperature.

Device junction temperature (T_j) can be calculated using $T_j = T_c + (P_{diss} \times R_{jc})$ where $P_{diss} = P_{dc} + P_{in} - P_{out}$ and T_c is the case temperature on the backside of the package (die pad) in contact with the PCB.

Ven Input

The device is enabled by applying a voltage between 1.8 V and 5.0 V to pin 16 (Ven). The resulting I_d taken by the device is relatively independent of the Ven voltage applied. If the enable feature is not required, the Ven pin can be connected to the same voltage as V_d .

The device can be placed into standby mode when not in use by setting Ven low (<0.2V) to disable all circuitry.

If lower I_{en} and I_d leakage current and/or if the highest forward isolation is needed in standby mode, a diode can be used in series with the Ven pin to increase the switch-on threshold of the device. This can be particularly important at elevated temperatures. Some suggested diodes in suitable packages (SOD-523F) for the evaluation board are:

- 1N914BWT-D PN fast switching diode
- BAT43XV2-D Schottky diode

Evaluation Board & Bias Procedure

In general, sequencing of the V_d and Ven supplies is not necessary however applying V_d before or simultaneously with Ven is recommended.

The separate through line can be used to measure the evaluation board and connector losses. These results can then be used to de-embed the device performance from evaluation board measurements.

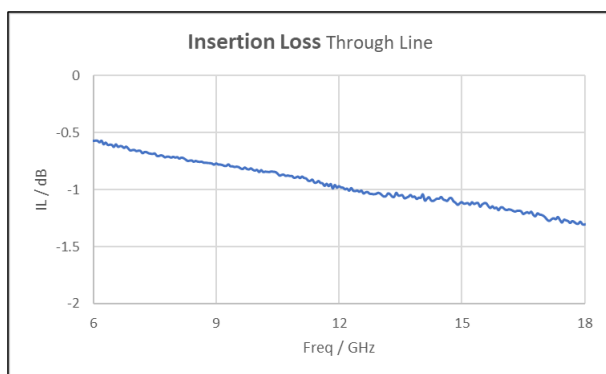


Figure 47: Insertion Loss – through line

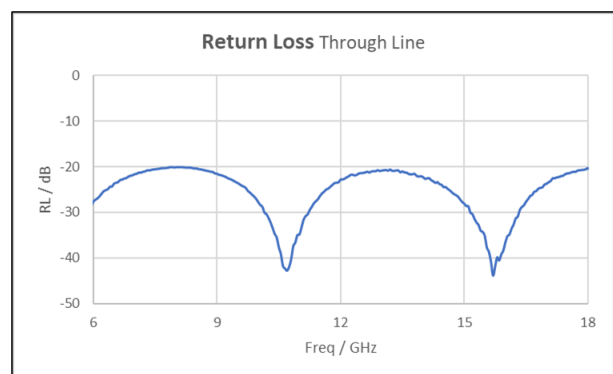
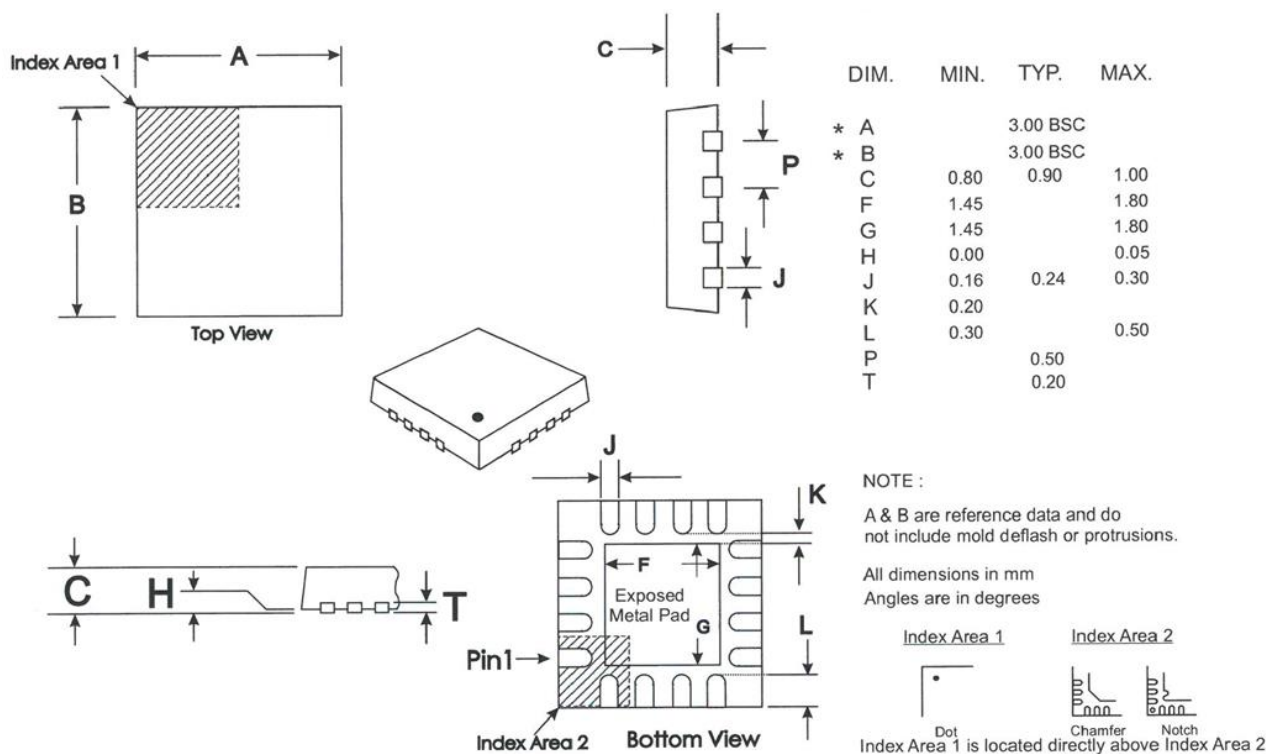


Figure 48: Return Loss – through line

Package Outline

16-lead 3x3mm VQFN Package (QF)



Package Marking

Pin 1 indicator (dot) and 3 rows of text for device identification.



Line 1: CMX90 SpμRF series
Line 2: 4-character part code
Line 3: Batch code

Revision History

Issue	Description	Date
1	First release for product launch	November 2022

Contact Information

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